



# FPGA Design Techniques

# Objectives

**After completing this module, you will be able to:**

- Increase design performance by duplicating flip-flops
- Increase design performance by adding pipeline stages
- Increase board performance by using I/O flip-flops
- Build reliable synchronization circuits



# Outline

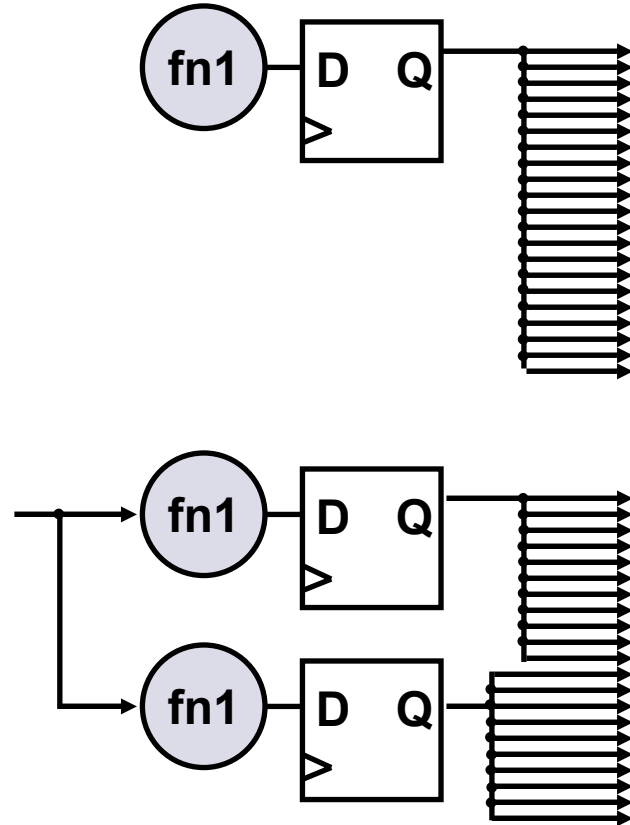


- **Duplicating Flip-flops**
- Pipelining
- I/O Flip-flops
- Synchronization Circuits
- Summary



# Duplicating Flip-Flops

- **High-fanout nets can be slow and hard to route**
- **Duplicating flip-flops can fix both problems**
  - Reduced fanout shortens net delays
  - Each flip-flop can fanout to a different physical region of the chip to help with routing
- **Design tradeoffs**
  - Gain routability and performance
  - Design area increases

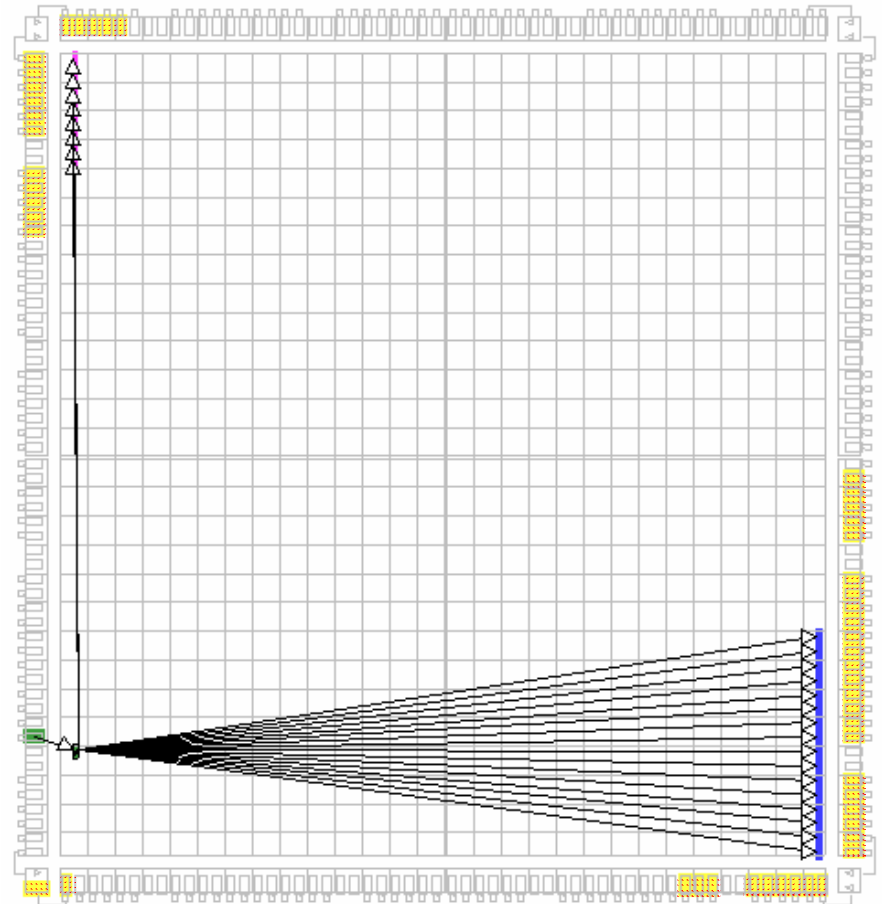


# Tips on Duplicating Flip-Flops

- **Name duplicated flip-flops `_a, _b`: *NOT* `_1, _2`**
  - Numbered flip-flops are mapped by default into the same slice
  - You generally want duplicated flip-flops to be separated
    - Especially if the loads are spread across the chip
- **Explicitly create duplicate flip-flops in your HDL code**
  - Most synthesis tools have automatic fanout-control features
    - However, they do not always pick the best division of loads
    - Also, duplicated flip-flops will be named `_1, _2`
  - Many synthesis tools will optimize-out duplicated flip-flops
    - Tell your synthesis tool to keep redundant logic
- **Do not duplicate flip-flops that are sourced by asynchronous signals**
  - Synchronize the signal first
  - Feed synchronized signal to multiple flip-flops

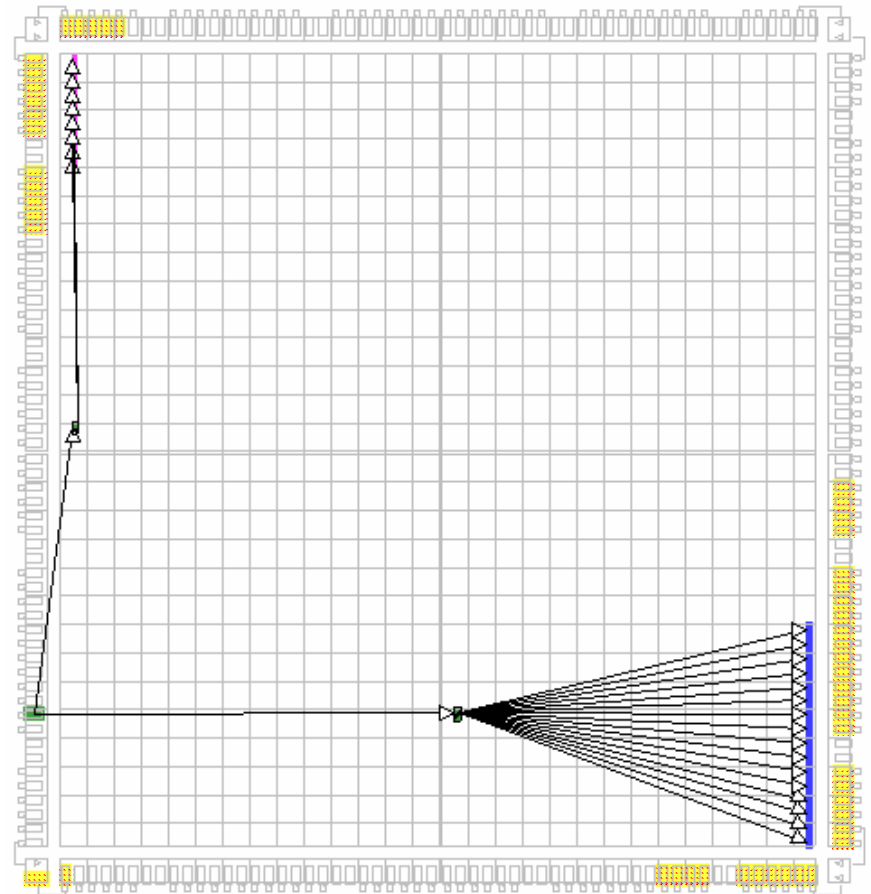
# Duplicating Flip-Flops Example

- Source flip-flop drives two register banks that are constrained to different regions of the chip
- Source flip-flop and pad are not constrained
- PERIOD = 5 ns timing constraint
- Implemented with default options
- Longest path = 6.806 ns
  - Fails to meet timing constraint



# Duplicating Flip-Flops Example

- **Source flip-flop has been duplicated**
- **Each flip-flop drives a region of the chip**
  - Each flip-flop can be placed closer to the register that it is driving
  - Shorter routing delays
- **Longest path = 4.666 ns**
  - Meets timing constraint



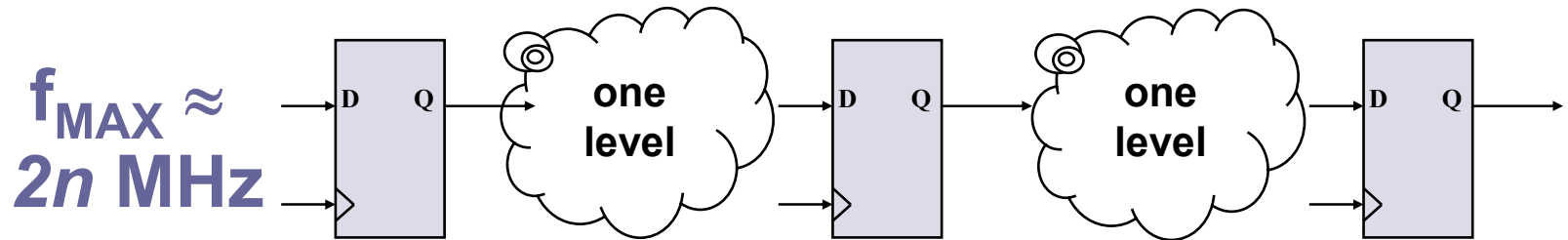
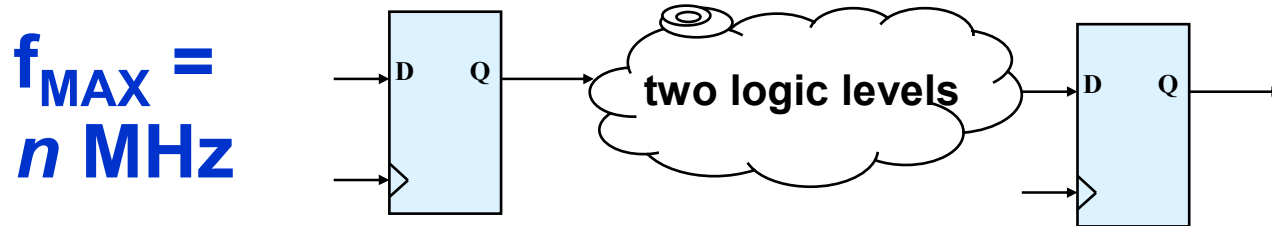
# Outline

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- **Pipelining**
- I/O Flip-flops
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# Pipelining Concept

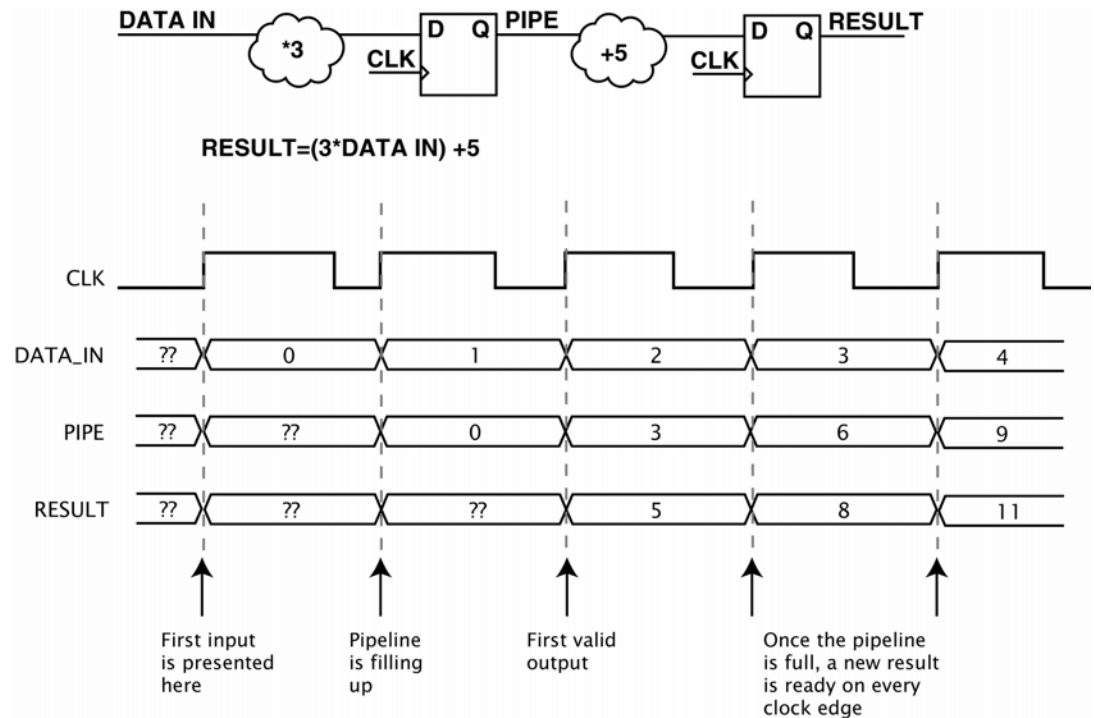


# Pipelining Considerations

- **Are there enough flip-flops available?**
  - Refer to the Map Report
  - You generally will not run out of FFS
- **Are there multiple logic levels between flip-flops?**
  - If there is only one logic level between flip-flops, pipelining will not improve performance
    - Exception: Long carry-logic chains can benefit from pipelining
  - Refer to the Post-Map Static Timing Report or Post-Place & Route Static Timing Report
- **Can the system tolerate *latency*?**

# Latency in Pipelines

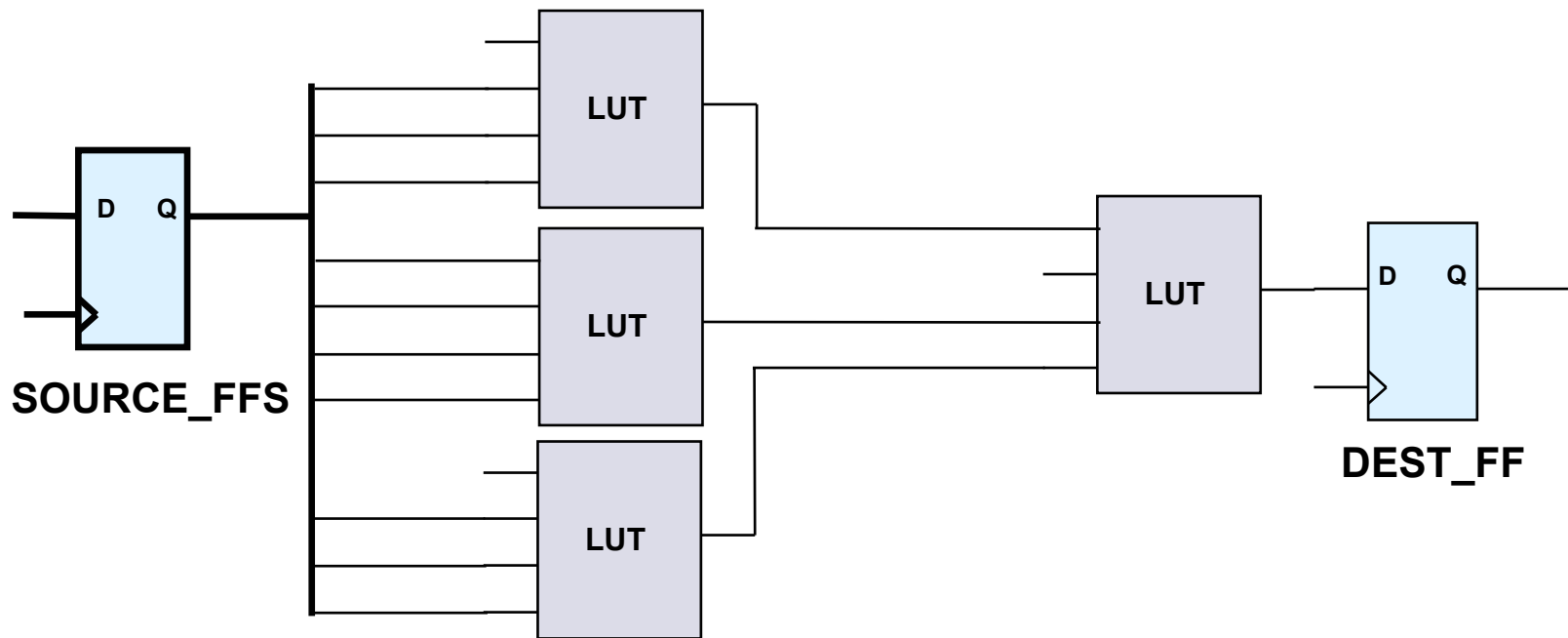
- Each pipeline stage adds one clock cycle of delay before the first output will be available
  - Also called “filling the pipeline”
- After the pipeline is filled, a new output is available every clock cycle



# Pipelining Example

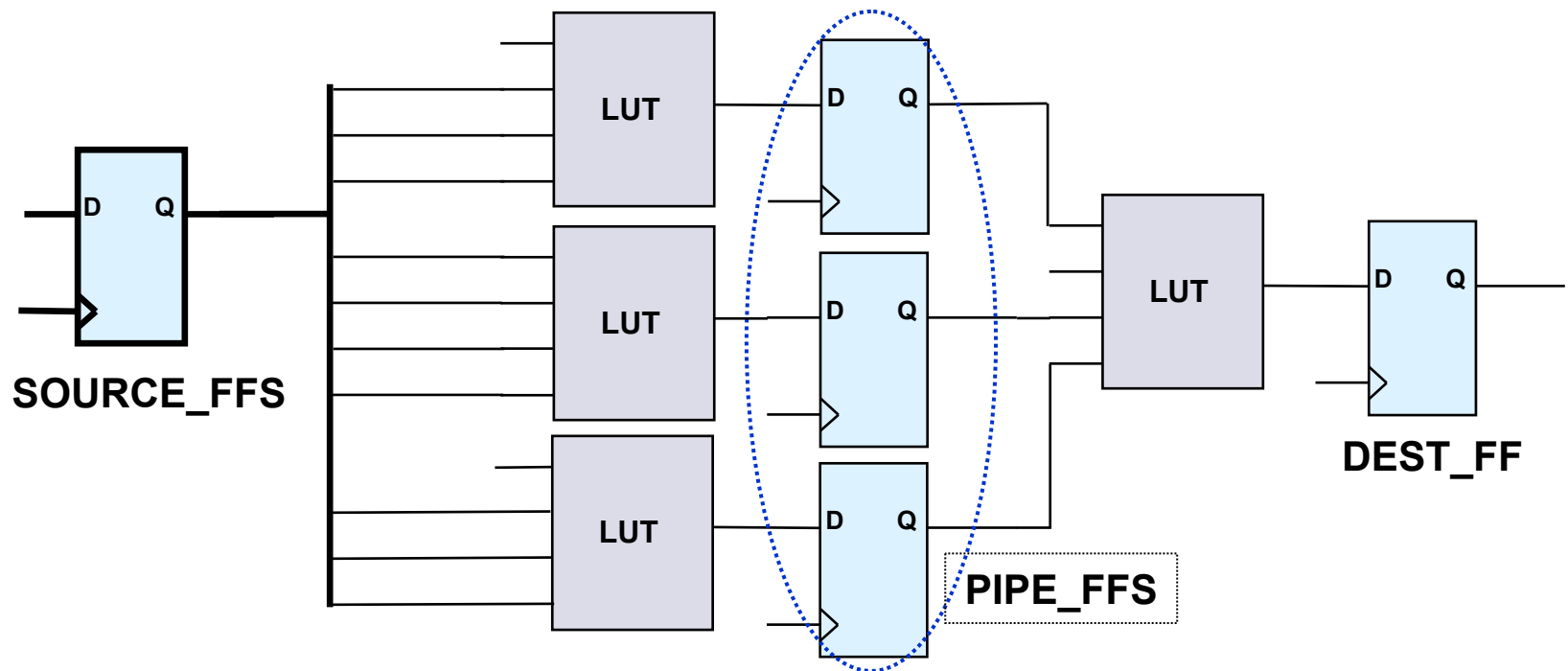
- Original circuit

- Two logic levels between SOURCE\_FFS and DEST\_FF
- $f_{MAX} = \sim 207$  MHz



# Pipelining Example

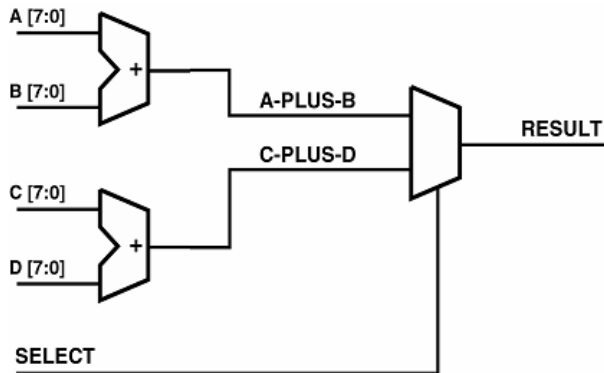
- **Pipelined circuit**
  - One logic level between each set of flip-flops
  - $f_{MAX} = \sim 347$  MHz



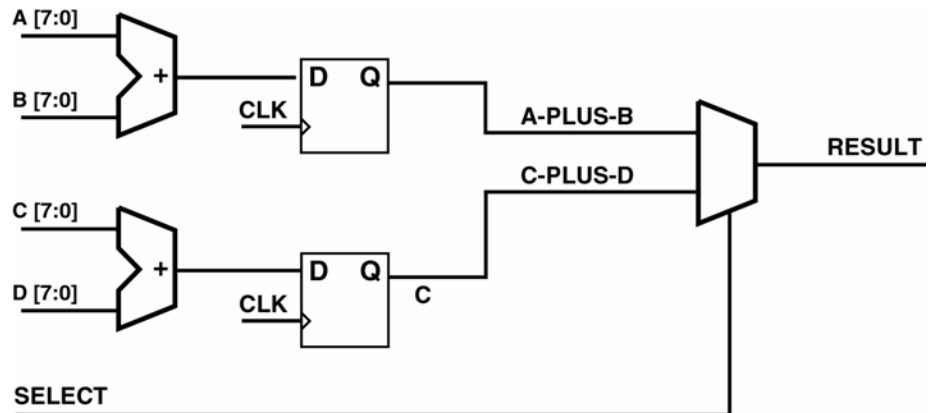
# Pipelining Questions

- Given the original circuit, what is wrong with the pipelined circuit?
- How can the problem be corrected?

Original Circuit

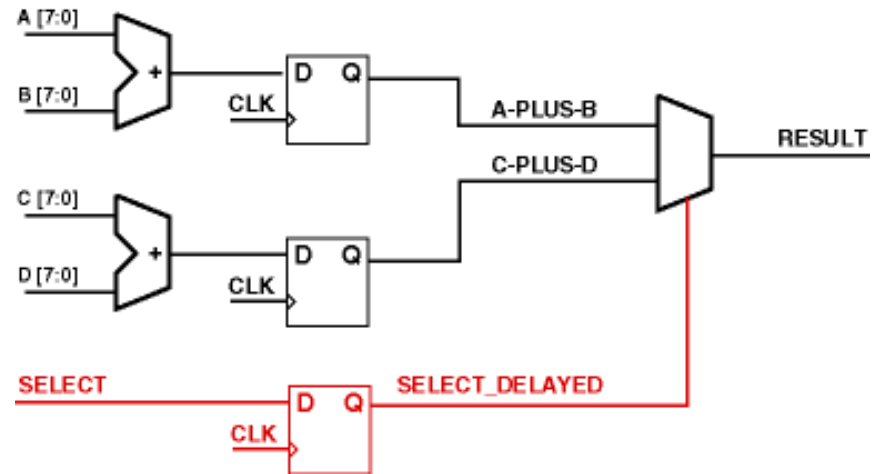


Pipelined Circuit



# Pipelining Answers

- **What is wrong with the pipelined circuit?**
  - Latency mismatch
  - Older data is mixed with newer data
  - Circuit output is incorrect
- **How can the problem be corrected?**
  - Add a flip-flop on *SELECT*
  - All data inputs now experience the same amount of latency



# Outline

- Duplicating Flip-Flops
- Pipelining
- **I/O Flip-flops**
- Synchronization Circuits
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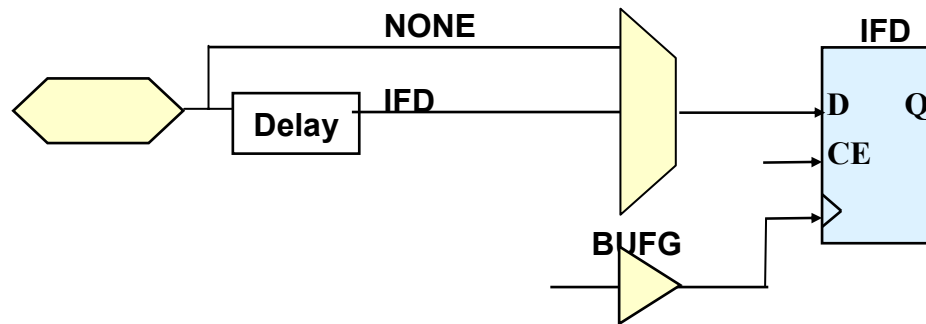
# I/O Flip-Flop Overview

- **Each Virtex™-II I/O Block contains six flip-flops**
  - IFD on the input, OFD on the output, ENBFF on the 3-state enable
  - Single or Double Data Rate support
- **Benefits of using I/O flip-flops**
  - Guaranteed setup, hold, and clock-to-out times when the clock signal comes from a BUFG
  - Programmable delay on input flip-flop
  - Programmable slew rate on output flip-flop



# Programmable Input Delay

- **Delay is added to the D input of the IFD or ILD to guarantee 0 ns hold time**
  - Trades an increase in setup time for a decrease in hold time
- **Delay is controlled by the IOBDELAY attribute**
  - NONE is the default if CLK is driven by a DCM
  - IFD is the default if CLK is not driven by a DCM



# Programmable Input Delay Example

- **XC2V250-5 (v1.90 Virtex™-II speed files)**
  - CLK is a DCM output (default delay is NONE)
    - IFD  $t_{SU} = 4.0$  ns  $t_H = 0.0$  ns
    - NONE  $t_{SU} = 1.3$  ns  $t_H = 0.0$  ns
  - CLK is not a DCM output (default delay is IFD)
    - IFD  $t_{SU} = 1.2$  ns  $t_H = 0.0$  ns
    - NONE  $t_{SU} = -1.5$  ns  $t_H = 1.7$  ns



# Programmable Output Slew Rate

- **Only the LVTTTL and LVCMOS I/O standards support programmable slew rate**
- **Slew rate can be SLOW or FAST**
  - By default, the slew rate is SLOW
- **Slew rate is controlled by attributes**
  - Specified in the Constraints Editor or the UCF file



# Accessing I/O Flip-Flops

- **During synthesis**
  - Timing-driven synthesis can force flip-flops into IOBs
  - Some tools support attributes or synthesis directives to mark flip-flops for placement in an IOB
- **Xilinx Constraint Editor**
  - Select the Misc tab
  - Specify individual registers that should be placed in the IOB
    - Need to know the instance name of each register
- **During the Map phase of implementation**
  - Map Properties, Pack I/O Registers/Latches into IOBs is on by default
  - Timing-driven packing will also move registers into IOBs for critical paths

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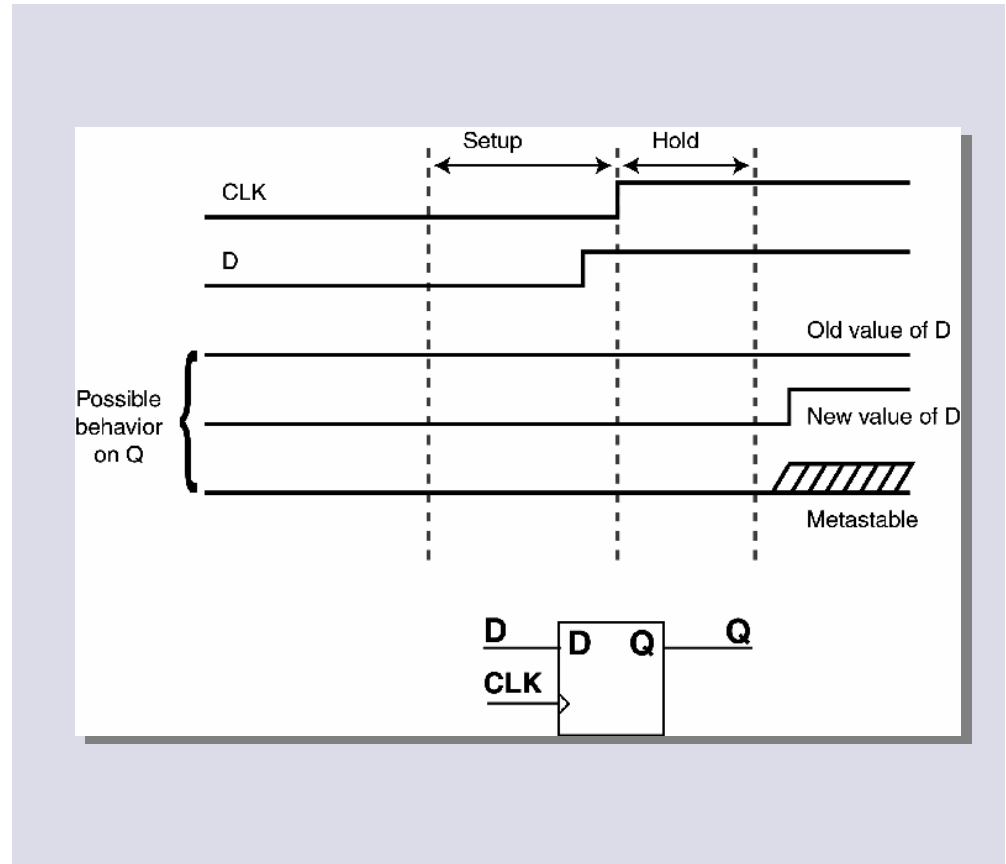


# Synchronization Circuits

- **What is a synchronization circuit?**
  - Captures an asynchronous input signal and outputs it on a clock edge
- **Why do I need synchronization circuits?**
  - Prevent setup and hold time violations
  - End result is a more reliable design
- **When do I need synchronization circuits?**
  - Signals cross between *unrelated* clock domains
    - Between related clock domains, a path-specific timing constraint is sufficient
  - Chip inputs that are asynchronous

# Setup and Hold Time Violations

- Violations occur when the flip-flop input changes too close to a clock edge
- Three possible results:
  - Flip-flop clocks in old data value
  - Flip-flop clocks in new data value
  - Flip-flop output becomes *metastable*





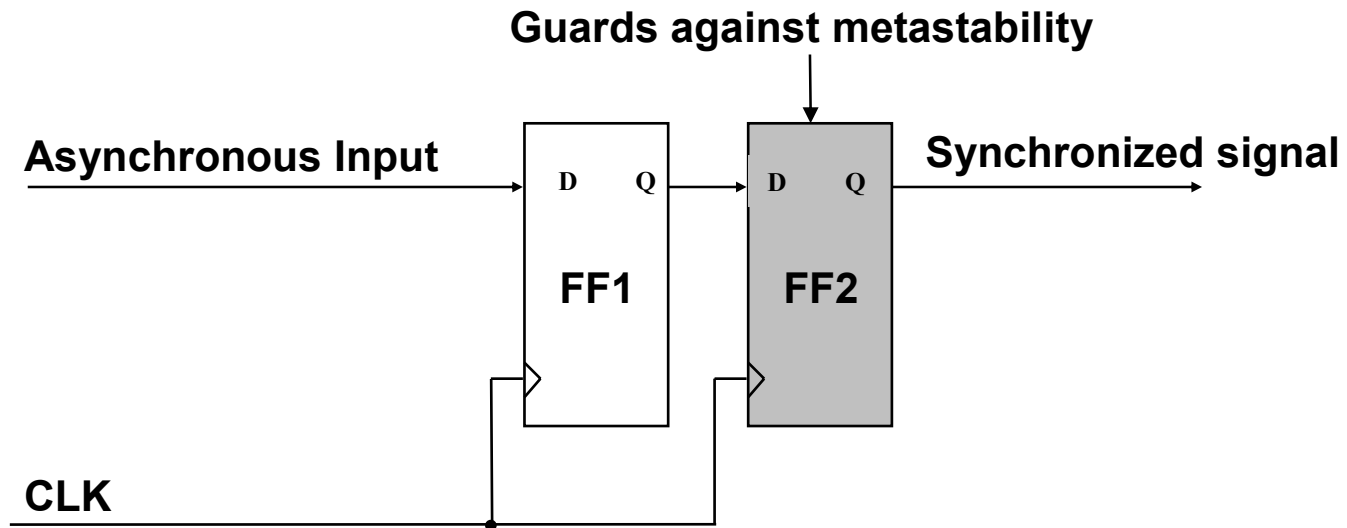
# Metastability

- **Flip-flop output enters a transitory state**
  - Neither a valid 0 nor a valid 1
    - May be interpreted as 0 by some loads and as 1 by others
  - Remains in this state for an unpredictable length of time before settling to a valid 0 or 1
- **Due to its statistical nature, the occurrence of metastable events can only be reduced, not eliminated**
- **Mean Time Between Failures (MTBF) is exponentially related to the length of time the flip-flop is given to recover**
  - A few extra ns of recovery time can dramatically reduce the chances of a metastable event
- **The circuits shown in this section allow a full clock cycle for metastable recovery**



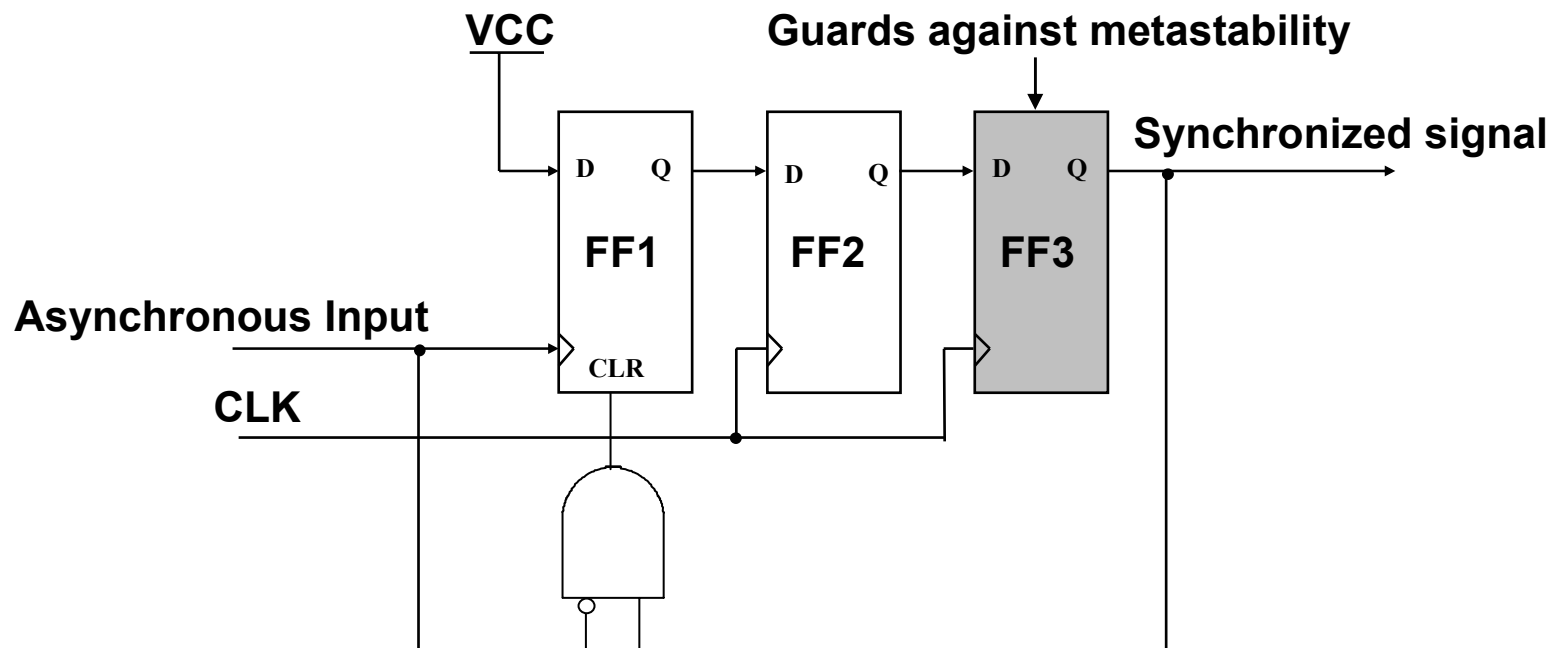
# Synchronization Circuit #1

- Use when input pulses will always be at least one clock period wide
- The “extra” flip-flops guard against metastability



# Synchronization Circuit #2

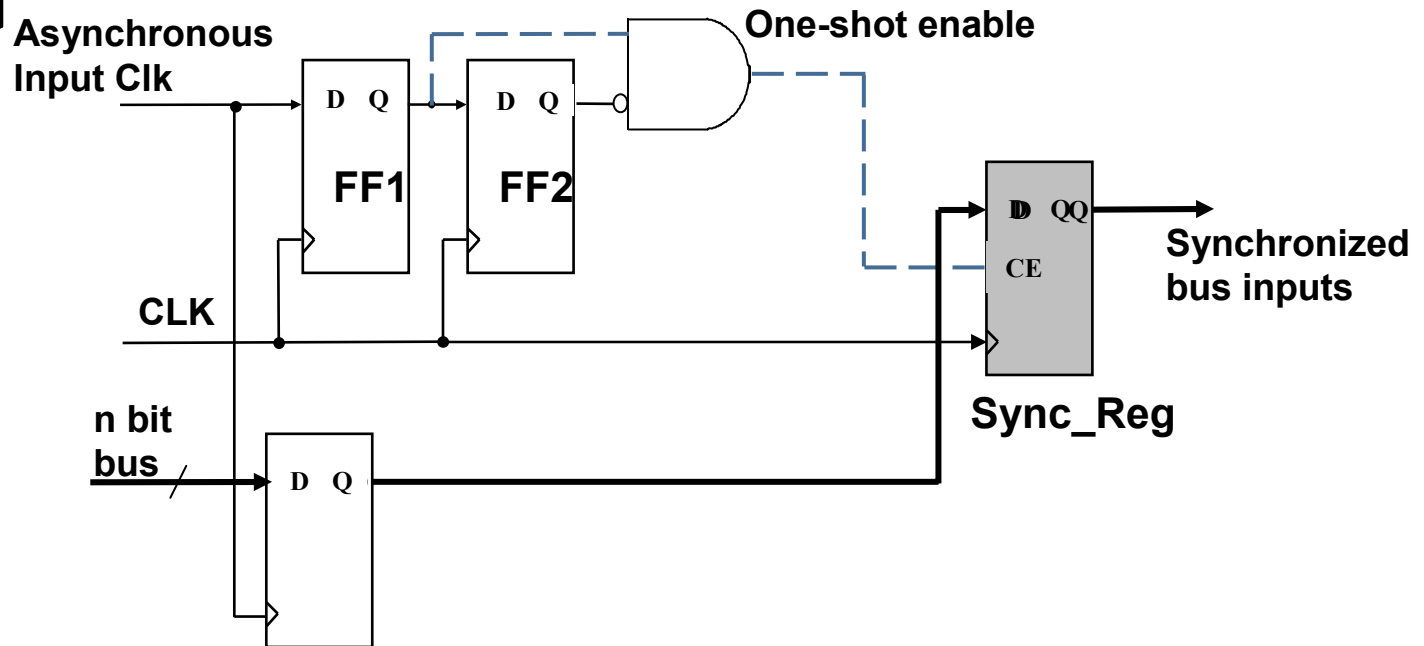
- Use when input pulses may be less than one clock period wide
  - FF1 captures short pulses



# Capturing a Bus

- **Leading edge detector**
  - Input pulses must be at least one CLK period wide

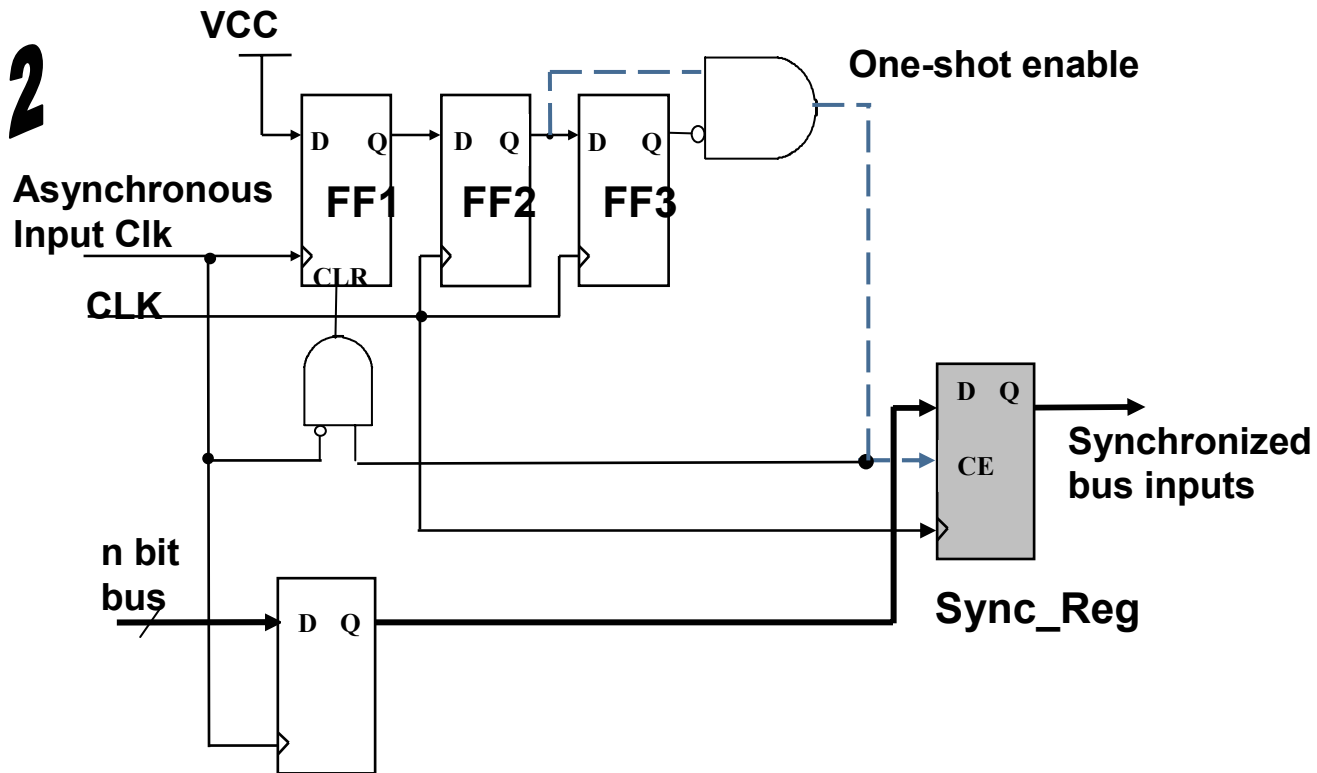
**Circuit 1**



# Capturing a Bus

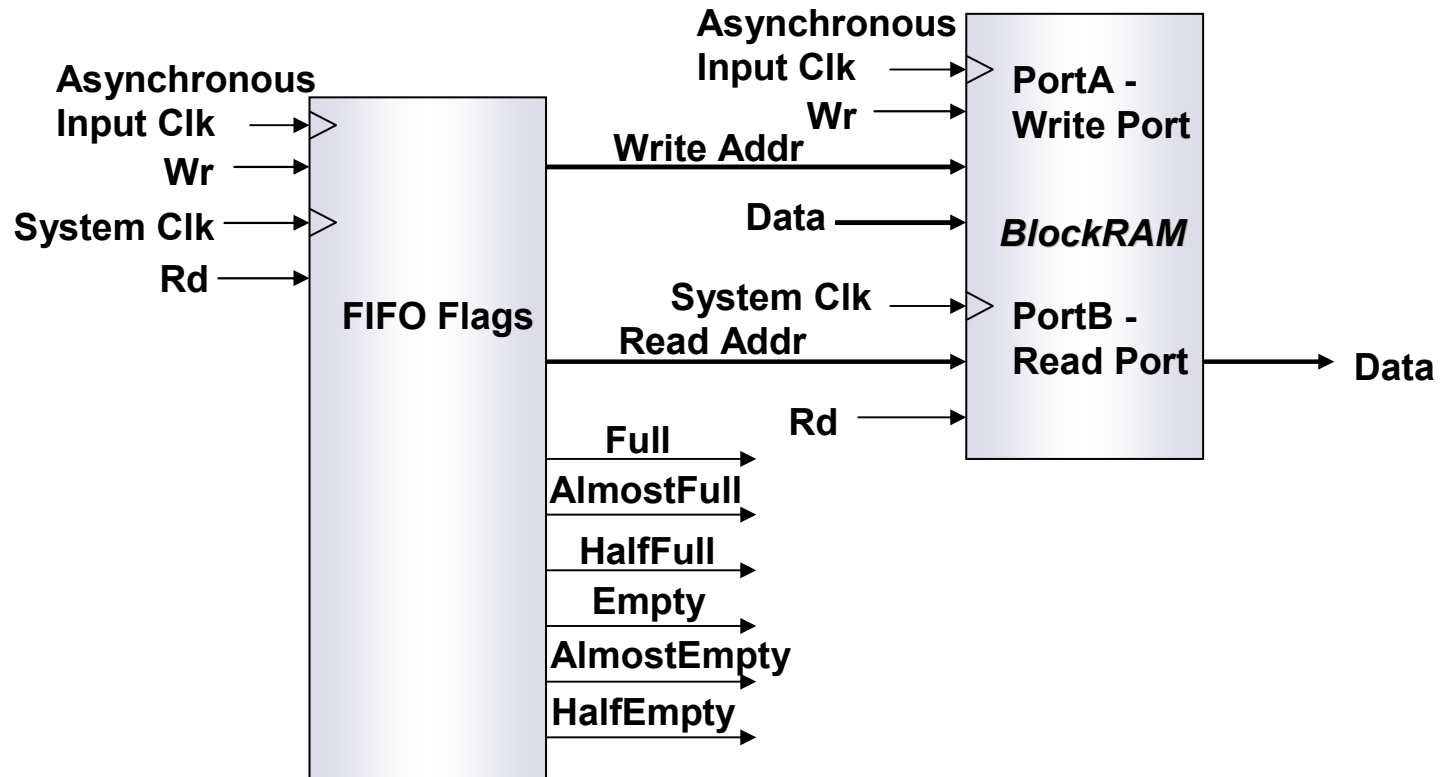
- **Leading edge detector**
  - Input pulses may be less than one CLK period wide

**Circuit 2**



# Synchronization Circuit #3

- Use a FIFO to cross domains



# Outline

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# Review Questions

- **High fanout is one reason to duplicate a flip-flop. What is another reason?**
- **Give an example of a time when you do not need to resynchronize a signal that crosses between clock domains**
- **What is the purpose of the “extra” flip-flop in the synchronization circuits shown in this module?**



# Answers

- **High fanout is one reason to duplicate a flip-flop. What is another reason?**
  - Loads are divided between multiple locations on the chip
- **Give an example of a time when you do not need to resynchronize a signal that crosses between clock domains**
  - Well-defined phase relationship between the clocks
  - Example: Clocks are the same frequency, 180° out of phase
  - Use a path-specific constraint to ensure that data paths will meet timing
- **What is the purpose of the “extra” flip-flop in the synchronization circuits shown in this module?**
  - To allow the first flip-flop time to recover from metastability

# Summary

- **You can increase circuit performance by:**
  - Duplicating flip-flops
  - Adding pipeline stages
  - Using I/O flip-flops
- **Some tradeoffs**
  - Duplicating flip-flops increases circuit area
  - Pipelining introduces latency
  - I/O flip-flop programmable delay trades off setup time and hold time
- **Synchronization circuits increase reliability**

# Where Can I Learn More?

- **Data Book: <http://support.xilinx.com> → Documentation → Databook**
  - Switching Characteristics
  - Detailed Functional Description > Input/Output Blocks (IOBs)
- **Application Notes: <http://support.xilinx.com> → Documentation → App Notes**
  - XAPP094: Metastability Recovery
  - XAPP225: Data to Clock Phase Alignment